

15. LTI System in Difference Equation

In system engineering, an n th-order discretized LTI system can be described by the following advanced form:

$$\begin{aligned} y[k+n] + a_{n-1}y[k+n-1] + \cdots + a_1y[k+1] + a_0y[k] \\ = b_nu[k+n] + b_{n-1}u[k+n-1] + \cdots + b_1u[k+1] + b_0u[k] \end{aligned} \quad (1)$$

which in fact is equivalent to the following delayed form

$$\begin{aligned} y[k] + a_{n-1}y[k-1] + \cdots + a_1y[k-n+1] + a_0y[k-n] \\ = b_nu[k] + b_{n-1}u[k-1] + \cdots + b_1u[k-n+1] + b_0u[k-n] \end{aligned} \quad (2)$$

In this course we will focus on the delayed form.

Assume $h[k]$ is the impulse response of a causal system and then $h[k]=0$ for $k<0$. It is known that for an initially relaxed system, i.e., $y[k]=u[k]=0$ for $k<0$, the input and output can be described by the following convolution

$$y[k] = h[k] * u[k] = \sum_{m=0}^k h[k-m]u[m] \quad (3)$$

which leads to

$$Y(z) = H(z)U(z) \quad (4)$$

where $H(z)$ is the transfer function of the system. According to (2), taking z-transform yields

$$\begin{aligned} Y(z) + a_{n-1}z^{-1}Y(z) + \cdots + a_1z^{-(n-1)}Y(z) + a_0z^{-n}Y(z) \\ = b_nU(z) + b_{n-1}z^{-1}U(z) + \cdots + b_1z^{-(n-1)}U(z) + b_0z^{-n}U(z) \end{aligned} \quad (5)$$

and thus

$$Y(z) = \frac{b_n + b_{n-1}z^{-1} + \cdots + b_1z^{-(n-1)} + b_0z^{-n}}{1 + a_{n-1}z^{-1} + \cdots + a_1z^{-(n-1)} + a_0z^{-n}} U(z) \quad (6)$$

Compared to (4), it is clear that the transfer function is

$$H(z) = \frac{Y(z)}{U(z)} = \frac{b_n + b_{n-1}z^{-1} + \cdots + b_1z^{-(n-1)} + b_0z^{-n}}{1 + a_{n-1}z^{-1} + \cdots + a_1z^{-(n-1)} + a_0z^{-n}} \quad (7)$$

which is in negative-power form. We can rewrite (7) into a positive-power form as

$$H(z) = \frac{b_n z^n + b_{n-1} z^{n-1} + \cdots + b_1 z + b_0}{z^n + a_{n-1} z^{n-1} + \cdots + a_1 z + a_0} \quad (8)$$

which is the form will be used in this course.

Similar to the Laplace transform, the discrete signals given as z-transform is also solved by using partial fraction expansion and then taking inverse z-transform. Here, we will adopt some examples for demonstration.

Example

What is the discrete signal $y[k]$ for $k \geq 0$ whose z-transform is

$$Y(z) = \frac{1}{(1 - z^{-1})(1 - 0.5z^{-1})}$$

Sol:

First, premultiply $Y(z)$ by z^2 and turn it into the positive-power form as

$$Y(z) = \frac{z^2}{(z-1)(z-0.5)}$$

Further express it as below:

$$\frac{Y(z)}{z} = \frac{z}{(z-1)(z-0.5)} = \frac{A_1}{z-1} + \frac{A_2}{z-0.5}$$

Hence, we have

$$z = A_1(z-0.5) + A_2(z-1)$$

and obtain $A_1 = 2$ and $A_2 = -1$. That results in

$$Y(z) = 2 \frac{z}{z-1} - 1 \frac{z}{z-0.5}$$

Taking the inverse z-transform, the discrete signal $y[k]$ is

$$y[k] = 2 - (0.5)^k \quad \text{for } k \geq 0.$$

Example

An initially relaxed LTI discrete system is given as

$$y[k] - y[k-1] + 0.5y[k-2] = u[k] + u[k-1]$$

- (A) What is the transfer function? Is the system stable?
- (B) What is the impulse response?
- (C) If $u[k]$ is a unit step input, then what is the output response $y[k]$?

Sol:

(A)

Taking the z-transform yields $Y(z) - z^{-1}Y(z) + 0.5z^{-2}Y(z) = U(z) + z^{-1}U(z)$, and then we obtain the transfer function as

$$\begin{aligned} H(z) &= \frac{Y(z)}{U(z)} = \frac{1+z^{-1}}{1-z^{-1}+0.5z^{-2}} = \frac{z^2+z}{z^2-z+0.5} \\ &= \frac{z+1}{(z-0.707e^{-j45^\circ})(z-0.707e^{j45^\circ})} \end{aligned}$$

where the two poles are located in the unit circle and then the system is stable.

(B)

Further express $H(z)$ as below:

$$\begin{aligned} \frac{H(z)}{z} &= \frac{z+1}{z^2-z+0.5} = \frac{z+1}{(z-0.5+j0.5)(z-0.5-j0.5)} \\ &= \frac{A}{z-0.707e^{-j45^\circ}} + \frac{A^*}{z-0.707e^{j45^\circ}} \end{aligned}$$

Hence, we have

$$z+1 = A(z-0.5-j0.5) + A^*(z-0.5+j0.5)$$

and obtain $A = 0.5 + j1.5 = \frac{\sqrt{10}}{2}e^{j71.6^\circ}$. That results in

$$H(z) = \frac{\sqrt{10}}{2}e^{j71.6^\circ} \frac{z}{z-0.707e^{-j45^\circ}} + \frac{\sqrt{10}}{2}e^{-j71.6^\circ} \frac{z}{z-0.707e^{j45^\circ}}$$

Taking the inverse z-transform, the impulse response $h[k]$ is

$$\begin{aligned} h[k] &= \frac{\sqrt{10}}{2}e^{j71.6^\circ} (0.707e^{-j45^\circ})^k + \frac{\sqrt{10}}{2}e^{-j71.6^\circ} (0.707e^{j45^\circ})^k \\ &= \frac{\sqrt{10}}{2} (0.707)^k e^{-j(45^\circ k - 71.6^\circ)} + \frac{\sqrt{10}}{2} (0.707)^k e^{j(45^\circ k - 71.6^\circ)} \\ &= \sqrt{10} (0.707)^k \cos(45^\circ k - 71.6^\circ) \end{aligned}$$

(C)

Since $U(z) = \frac{z}{z-1}$, we have

$$\begin{aligned} Y(z) &= H(z)U(z) = \frac{z(z^2+z)}{(z-1)(z^2-z+0.5)} \\ &= \frac{A_1 z}{z-1} + \frac{A_2 z}{z-0.707e^{-j45^\circ}} + \frac{A_2^* z}{z-0.707e^{j45^\circ}} \end{aligned}$$

where $A_1 = 4$ and $A_2 = 1.58e^{j161.6^\circ}$. Hence, we have

$$y[k] = 4 + 3.16(0.707)^k \cos(45^\circ k - 161.6^\circ)$$

Exercise:

An initially relaxed LTI discrete system is given as

$$y[k] + 0.1y[k-1] + 0.2y[k-2] = u[k] + u[k-1]$$

- (A) What is the transfer function? Is the system stable?
- (B) What is the impulse response?
- (C) If $u[k]$ is a unit step input, then what is the output response $y[k]$?