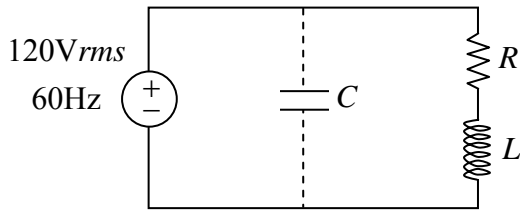


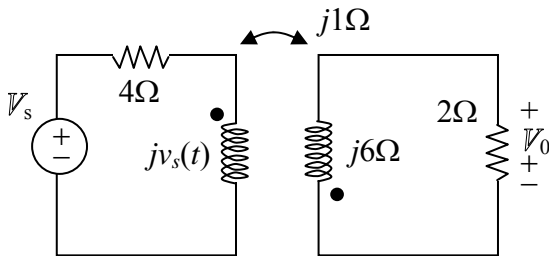
1. Consider the following circuit with the load $R=10\Omega$ and $L=0.1\text{H}$ in series.



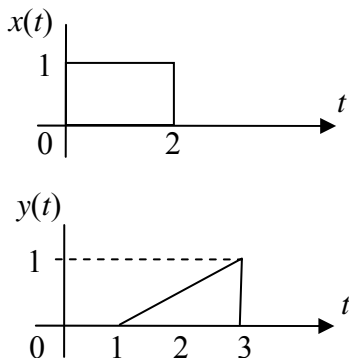
- [6%] (a) What is the complex power delivered by the source?
 [3%] (b) What is the average power dissipated?
 [3%] (c) What is the power factor of the load?
 [6%] (d) Choose the capacitance C that will give a unity power factor?

2. In a balanced three phase Y-Y system, the source is an abc sequence of voltages and $V_{an}=120 \angle 30^\circ \text{V}_{rms}$. The line impedance per phase is $0.5+j1 \Omega$, while the per-phase impedance of the load is $10+j15 \Omega$. Calculate the three load voltages.

3. Let $V_s=10 \angle 60^\circ$. What is the phasor voltage V_0 ?

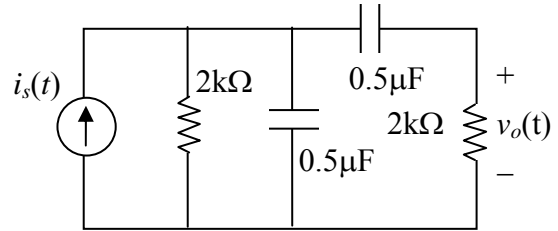


4. Draw $z(t) = x(t) * y(t)$, the integral convolution of $x(t)$ and $y(t)$



5. Neglect the initial conditions of the capacitors in the following circuit with input $i_s(t)$ and output $v_o(t)$.

- [10%] (a) Determine the transfer function $H(s)$.
 [6%] (b) Find its poles and zeros.
 [8%] (c) If the output is $v_o(t)=V_o \cos(\omega t - \pi/4)$ when $i_s(t)=\cos \omega t$, what are the amplitude V_o and the frequency ω ?



6. Determine the inverse Laplace transform of the following signals:

[4%] (a) $F(s) = \frac{3s+1}{(s+1)(s+4)}$

[5%] (b) $G(s) = \frac{s-1}{s(s+1)^2}$

[5%] (c) $H(s) = \frac{2s^2-3}{(s+2)(s^2+2s+5)}$

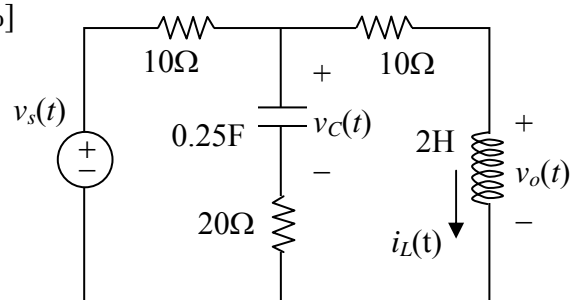
7. Let $P(s) = \frac{s^2+3}{s^4-3s^2+2s}$ and determine the

- [5%] initial and final values of $p(t)$.

8. Consider the following circuit with a dc voltage source $v_s(t)=10 \text{V}$ and initial conditions $v_C(0^-)=1 \text{V}$ and $i_L(0^-)=0 \text{A}$.

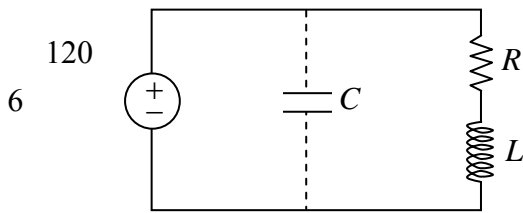
- (a) Choose $v_C(t)$ and $i_L(t)$ as the state variables and write the state equation.
 [15%] (b) Determine the output voltage $v_o(t)$.

[15%]



Solution

1. Consider the following circuit with the load $R=10\Omega$ and $L=0.1\text{H}$ in series.



- What is the complex power delivered by the source?
- What is the average power dissipated?
- What is the power factor of the load?
- Choose the capacitance C that will give a unity power factor?

Solution-01

The impedance of the load is

$$\begin{aligned} Z &= R + j\omega L = 10 + j60 \times 0.1 \\ &= 10 + j6 = 11.66 \angle 30.96^\circ \end{aligned}$$

The current through the load is

$$\begin{aligned} I &= 120 \angle 0^\circ / Z = 120 \angle 0^\circ / 11.66 \angle 30.96^\circ \\ &= 8.8235 - j5.2941 = 10.29 \angle -30.96^\circ \end{aligned}$$

- (a) The complex power

$$\begin{aligned} S &= VI^* = 120 \angle 0^\circ \times 10.29 \angle 30.96^\circ \\ &= 1234.8 \angle 30.96^\circ = 1058.8 + j635.3 \text{ (VA)} \end{aligned}$$

- (b) The average power is 1058.8 W.
(c) The power factor is

$$pf = \cos\left(\frac{1058.8}{1234.8}\right) = 0.6544$$

- (d) The complex power in C is

$$\begin{aligned} S &= VI^* = V(j\omega CV)^* \\ &= -j \times 60 \times 120^2 C \end{aligned}$$

To correct the power factor, we let

$$60 \times 120^2 C = 635.3 \Rightarrow C = 735$$

Hence, $C=735\mu\text{F}$.

2. In a balanced three phase Y-Y system, the

source is an abc sequence of voltages and $V_{an}=120 \angle 30^\circ \text{ V}_{rms}$. The line impedance per phase is $0.5+j1 \Omega$, while the per-phase impedance of the load is $10+j15 \Omega$. Calculate the three load voltages.

Solution-02

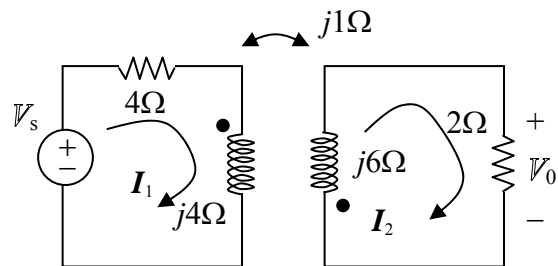
The load voltage V_{An} is

$$\begin{aligned} V_{An} &= \frac{10 + j15}{0.5 + j1 + 10 + j15} V_{an} \\ &= (0.9420 - j0.0068) \times 120 \angle 30^\circ \\ &= 0.9420 \angle -0.415^\circ \times 120 \angle 30^\circ \\ &= 113.04 \angle 29.585^\circ \end{aligned}$$

Then the other two load voltages are

$$\begin{aligned} V_{Bn} &= 113.04 \angle -90.415^\circ \\ V_{Cn} &= 113.04 \angle 149.585^\circ \end{aligned}$$

3. Let $V_s=10 \angle 60^\circ$. What is the phasor voltage V_0 ?



Solution-03

The mesh equations are

$$\begin{aligned} V_s &= (4 + j4)I_1 + jI_2 \\ 0 &= jI_1 + (2 + j6)I_2 \end{aligned}$$

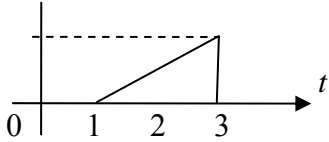
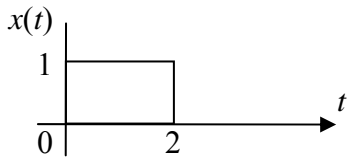
Hence, $I_1 = j(2 + j6)I_2 = (-6 + j2)I_2$. Then

$$\begin{aligned} V_s &= (4 + j4)(-6 + j2)I_2 + jI_2 \\ &= (-32 - j15)I_2 \end{aligned}$$

It results in $I_2 = -\frac{V_s}{32 + j15}$. Therefore,

$$\begin{aligned} V_0 &= 2I_2 = -\frac{2V_s}{32 + j15} \\ &= -\frac{2 \times 10 \angle 60^\circ}{32 + j15} = 0.5659 \angle -145.1^\circ \end{aligned}$$

4. Draw $z(t) = x(t) * y(t)$, the integral convolution of $x(t)$ and $y(t)$



Solution-04

The integral convolution is defined as

$$z(t) = x(t) * y(t) = \int_{-\infty}^{\infty} x(\tau) y(t-\tau) d\tau$$

In this case, we have

$$z(t) = x(t) * y(t) = \int_0^t x(\tau) y(t-\tau) d\tau \quad \text{for } t > 0.$$

For $0 < t < 1$, $z(t) = 0$.

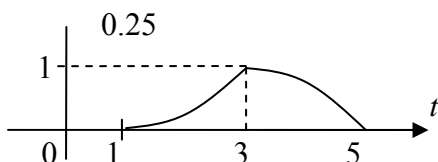
For $1 < t < 3$,

$$\begin{aligned} z(t) &= \int_0^{t-1} 1 \cdot \left(\frac{t-1-\tau}{2} \right) d\tau \\ &= \left. \frac{(t-1)\tau - 0.5\tau^2}{2} \right|_{\tau=0}^{t-1} = \frac{(t-1)^2}{4} \end{aligned}$$

For $3 < t < 5$,

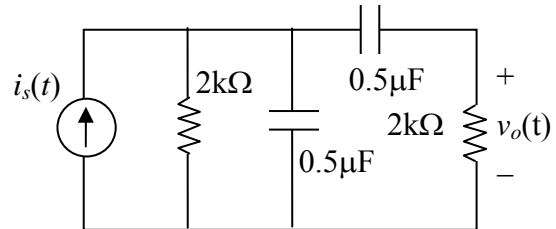
$$\begin{aligned} z(t) &= \int_{t-3}^2 1 \cdot \left(\frac{t-1-\tau}{2} \right) d\tau \\ &= \left. \frac{(t-1)\tau - 0.5\tau^2}{2} \right|_{\tau=t-3}^2 \\ &= -0.25t^2 + 1.5t - 1.25 \end{aligned}$$

For $t > 5$, $z(t) = 0$.



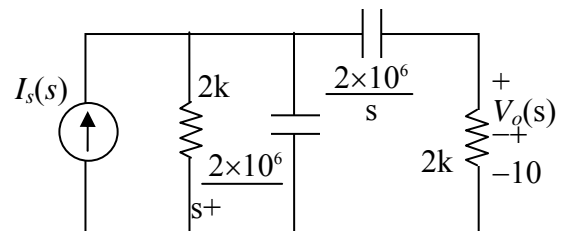
5. Neglect the initial conditions of the capacitors in the following circuit with input $i_s(t)$ and output $v_o(t)$.

- Determine the transfer function $H(s)$.
- Find its poles and zeros.
- If the output is $v_o(t) = V_o \cos(\omega t - \pi/4)$ when $i_s(t) = \cos \omega t$, what are the amplitude V_o and the frequency ω ?



Solution-05

In s-form, we redraw the circuit as below:



- Let $G = 1/(2k)$ and $C = 0.5\mu$, then the output voltage is

$$V_o = \frac{1}{G} \cdot \frac{\frac{1}{sC} + \frac{1}{G}}{G + sC + \frac{1}{\frac{1}{sC} + \frac{1}{G}}} \cdot I_s$$

Hence,

$$\begin{aligned} H(s) &= \frac{V_o}{I_s} = \frac{sC}{(G + sC)^2 + sCG} \\ &= \frac{2 \times 10^6 s}{s^2 + 3000s + 10^6} \end{aligned}$$

- Clearly, there is one zero at $s=0$. Solve $s^2 + 3000s + 10^6 = 0$ and obtain the two poles at $s = -2618$ and -382 .

- Since $H(j\omega) = \frac{j\omega 2 \times 10^6}{G(j\omega)}$, where

$$G(j\omega) = (j\omega)^2 + 3000(j\omega) + 10^6$$

$$= 10^6 - \omega^2 + j3000\omega$$

the phase of $H(j\omega)$

is $90^\circ - \angle G(j\omega) = -45^\circ$.

Hence, $\angle G(j\omega) = 135^\circ$,

i.e., $3000\omega = \omega^2 - 10^6$.

Then, $\omega = 3302.8$ rad/s.

Thus,

$$V_o = |H(j3302.8)| = 471.4024.$$

6. Determine the inverse Laplace transform of the following signals:

$$(a) F(s) = \frac{3s+1}{(s+1)(s+4)}$$

$$(b) G(s) = \frac{s-1}{s(s+1)^2}$$

$$(c) H(s) = \frac{2s^2-3}{(s+2)(s^2+2s+5)}$$

Solution-06

(a)

$$F(s) = \frac{3s+1}{(s+1)(s+4)} = -\frac{2/3}{s+1} + \frac{11/3}{s+4}$$

$$\text{Hence, } f(t) = -\frac{2}{3}e^{-t} + \frac{11}{3}e^{-4t} \text{ for } t > 0.$$

(b)

$$G(s) = \frac{s-1}{s(s+1)^2} = \frac{-1}{s} + \frac{1}{s+1} + \frac{2}{(s+1)^2}$$

$$\text{Hence, } g(t) = -1 + e^{-t} + 2te^{-t} \text{ for } t > 0.$$

(c)

$$H(s) = \frac{2s^2-3}{(s+2)(s^2+2s+5)}$$

$$= \frac{1}{s+2} + \frac{(s+1) + 2(-5/2)}{(s+1)^2 + 2^2}$$

$$\text{Hence, } h(t) = e^{-2t} + e^{-t} \left(\cos 2t - \frac{5}{2} \sin 2t \right)$$

for $t > 0$.

7. Let $P(s) = \frac{s^2+3}{s^4-3s^2+2s}$ and determine the initial and final values of $p(t)$.

Solution-07

For the initial value, we have

$$p(0) = \lim_{s \rightarrow \infty} sP(s) = \lim_{s \rightarrow \infty} \frac{s(s^2+3)}{s^4-3s^2+2s}$$

$$= \lim_{s \rightarrow \infty} \frac{s^2+3}{s^3-3s+2} = 0$$

Since there exist poles located in the right half complex plane, it is unbounded for $t \rightarrow \infty$ and the final value theorem is not suitable to obtain $p(\infty)$. It must be solved as below:

$$P(s) = \frac{s^2+3}{s^4-3s^2+2s} = \frac{s^2+3}{s(s+2)(s-1)^2}$$

$$= \frac{A}{s} + \frac{B}{s+2} + \frac{C}{s-1} + \frac{4/3}{(s-1)^2}$$

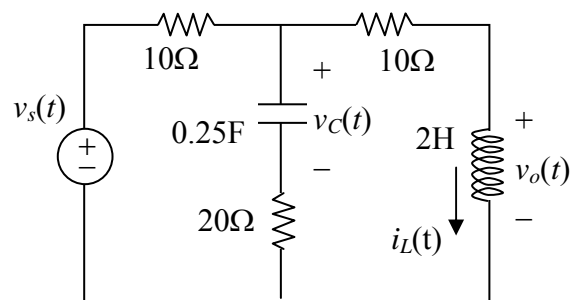
Hence, $p(t) = A + Be^{-2t} + Ce^t + \frac{4}{3}te^t$ for $t > 0$.

Thus, $p(\infty) = \infty$.

8. Consider the following circuit with a dc voltage source $v_s(t) = 10$ V and initial conditions $v_C(0^-) = 1$ V and $i_L(0^-) = 0$ A.

(a) Choose $v_C(t)$ and $i_L(t)$ as the state variables and write the state equation.

(b) Determine the output voltage $v_o(t)$.



Solution-08

(a) With mesh analysis, the mesh current equation for mesh 1 is

$$10i_1(t) + v_C(t) + 20(i_1(t) - i_2(t)) = v_s(t)$$

and the mesh current equation for mesh 2 is

$$i_2(t) = i_L(t)$$

Clearly, replacing $i_2(t)$ with $i_L(t)$ for mesh 1 yields

$$i_1(t) = -\frac{1}{30}v_c(t) + \frac{2}{3}i_L(t) + \frac{1}{30}v_s(t)$$

The component equation of the capacitor is

$$0.25 \frac{dv_c(t)}{dt} = i_1(t) - i_2(t)$$

\Rightarrow

$$\begin{aligned} \frac{dv_c(t)}{dt} &= 4(i_1(t) - i_2(t)) \\ &= 4\left(-\frac{1}{30}v_c(t) + \frac{2}{3}i_L(t) + \frac{1}{30}v_s(t) - i_2(t)\right) \\ &= -\frac{2}{15}v_c(t) - \frac{4}{3}i_L(t) + \frac{2}{15}v_s(t) \end{aligned}$$

The component equation of the inductor is

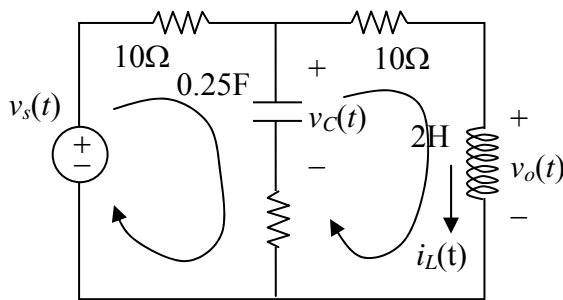
$$2 \frac{di_L(t)}{dt} = v_s(t) - 10i_1(t) - 10i_2(t)$$

\Rightarrow

$$\begin{aligned} \frac{di_L(t)}{dt} &= 0.5v_s(t) - 5i_1(t) - 5i_2(t) \\ &= 0.5v_s(t) - 5\left(-\frac{1}{30}v_c(t) + \frac{2}{3}i_L(t) + \frac{1}{30}v_s(t) + i_L(t)\right) \\ &= \frac{1}{6}v_c(t) - \frac{25}{3}i_L(t) + \frac{1}{3}v_s(t) \end{aligned}$$

Hence, the state equation in matrix form is

$$\begin{bmatrix} \frac{dv_c(t)}{dt} \\ \frac{di_L(t)}{dt} \end{bmatrix} = \begin{bmatrix} -\frac{2}{15} & -\frac{4}{3} \\ \frac{1}{6} & -\frac{25}{3} \end{bmatrix} \begin{bmatrix} v_c(t) \\ i_L(t) \end{bmatrix} + \begin{bmatrix} \frac{2}{15} \\ \frac{1}{3} \end{bmatrix} v_s(t)$$



(b) Taking Laplace transform of the state equation, we have

$$\begin{aligned} s \begin{bmatrix} V_c(s) \\ I_L(s) \end{bmatrix} - \begin{bmatrix} 1 \\ 0 \end{bmatrix} \\ = \begin{bmatrix} -\frac{2}{15} & -\frac{4}{3} \\ \frac{1}{6} & -\frac{25}{3} \end{bmatrix} \begin{bmatrix} V_c(s) \\ I_L(s) \end{bmatrix} + \begin{bmatrix} \frac{2}{15} \\ \frac{1}{3} \end{bmatrix} \frac{10}{s} \end{aligned}$$

i.e.,

$$\begin{bmatrix} s + \frac{2}{15} & \frac{4}{3} \\ -\frac{1}{6} & s + \frac{25}{3} \end{bmatrix} \begin{bmatrix} V_c(s) \\ I_L(s) \end{bmatrix} = \begin{bmatrix} \frac{2}{15} \\ \frac{1}{3} \end{bmatrix} \frac{10}{s} + \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$\Rightarrow \begin{bmatrix} V_c(s) \\ I_L(s) \end{bmatrix} = \begin{bmatrix} s + \frac{2}{15} & \frac{4}{3} \\ -\frac{1}{6} & s + \frac{25}{3} \end{bmatrix}^{-1} \begin{bmatrix} 1 + \frac{20}{15s} \\ \frac{10}{3s} \end{bmatrix}$$

$$\Rightarrow \begin{bmatrix} V_c(s) \\ I_L(s) \end{bmatrix} = \frac{\begin{bmatrix} s + \frac{25}{3} & -\frac{4}{3} \\ \frac{1}{6} & s + \frac{2}{15} \end{bmatrix} \begin{bmatrix} 1 + \frac{20}{15s} \\ \frac{10}{3s} \end{bmatrix}}{s^2 + \frac{127}{15}s + \frac{4}{3}}$$

$$\begin{aligned} \Rightarrow I_L(s) &= \frac{\frac{7}{2} + \frac{2}{3s}}{s^2 + \frac{127}{15}s + \frac{4}{3}} \\ &= \frac{0.5}{s} + \frac{-0.4198}{s + 8.3061} + \frac{-0.0803}{s + 0.1605} \end{aligned}$$

Taking the inverse Laplace transform, we obtain

$$i_L(t) = 0.5 - 0.4198e^{-8.3061t} - 0.0803e^{-0.1605t}$$

Hence, the voltage

$$\begin{aligned} v_o(t) &= 2 \frac{di_L(t)}{dt} \\ &= 6.9738e^{-8.3061t} + 0.0258e^{-0.1605t} \end{aligned}$$